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Hybrid photovoltaic-thermal solar collectors dynamic modeling

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HIGHLIGHTS

- ► A hybrid photovoltaic/thermal dynamic model is presented.
- ▶ The model, once calibrated, can predict the power output for any set of climate data.
- ▶ The physical electrical model includes explicitly thermal and irradiance dependences.
- ▶ The results agree with those obtained through steady-state characterization.
- ► The model approaches the junction cell temperature through the system energy balance.

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ABSTRACT

A hybrid photovoltaic/thermal transient model has been developed and validated experimentally. The methodology extends the quasi-dynamic thermal model stated in the EN 12975 in order to involve the electrical performance and consider the dynamic behavior minimizing constraints when characterizing the collector. A backward moving average filtering procedure has been applied to improve the model response for variable working conditions. Concerning the electrical part, the model includes the thermal and radiation dependences in its variables. The results revealed that the characteristic parameters included in the model agree reasonably well with the experimental values obtained from the standard steady-state and IV characteristic curve measurements. After a calibration process, the model is a suitable tool to predict the thermal and electrical performance of a hybrid solar collector, for a specific weather data set.

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1. Introduction

Withi, blar energy technologies, the hybrid photovoltaic-thermal (PV/T) systems offer an attractive option because the absorbed solar radiation is converted into thermal and electrical energy (the conversion can be done separately or simultaneously). Flat-plate collectors are the most common devices that combine thermal and electrical generation. In order to evaluate these types of collectors, a performance analysis must be undertaken in accordance with the standard test methods available. This can provide objective information, valuable for designers and manufactures in the frame of collector design and system performance optimization for a particular application. Moreover, consumers can also compare the performance and cost-effectiveness ratio of a competing product. Albeit, no dedicated guideline, as a standard test procedure, exists for the performance testing of PV/T flat-plate collectors until now. This field has not been investigated extensively in the litera-

* Corresponding author. Tel.: +34 973003711; fax: +3¹⁸ 3003575. *E-mail address:* daniel.chemisana@macs.udl.cat (D. Chemisana). ture. The area references refer to PV/T models that can be applied to transient weather conditions.

On the other hand, a wide number of studies associated with the PV/T collector. Lave been reported in the literature. Most of them are studies about electrical and thermal performances under steady-state conditions [1–3]. Nevertheless, steady-state outdoor testing includes several difficulties associated with the weather conditions. In many places in the world and over many periods throughout the year, the weather conditions do not fulfil the requirements for the steady-state method and long testing periods may be needed to acquire steady-state values. Dynamic or quasidynamic characterization methods can offer an effective alternative to overcome these problems. Some representative references regarding quasi-dynamic and dynamic procedures are described below.

Bernardo et al. [4] proposed an empirical model which approaches well the specific thermal power under unstable conditions. However, the specific electrical power predicted by the model showed wider divergences in comparison with the results shown in the thermal part. This system works under concentrated





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Nomenclature

A _{col}	rea of the collector/module (m^2)
<i>c</i> ₁	Leat loss coefficient at $(T_m - T_a) = 0$ (W/(m ² K))
<i>c</i> ₂	temperature dependence of the heat loss coefficient (W/
	$(m^2 K^2))$
<i>c</i> ₅	effective thermal capacity $(J/(m^2 K))$
c_p	²⁴ pecific heat of fluid (J/kg K)
d ₁	radiation dependence of the photogenerated current
	$(Am^2 W^{-1})$
d_2	temperature dependence of the saturation diode current
	(AK^{-3})
d ₃	shunt conductance (A/V)
E_g	band gap energy (J)
F'	collector efficiency factor (–)
FF	fill factor of PV/T module (–)
F' (τα) _{en}	zero loss efficiency (–)
f_c	cut frequency (s ⁻¹)
f_N	Nyquist frequency (s ⁻¹)
G	solar radiation (W/m^2)
G^*	thermal effective solar radiation (W/m^2)
Iph	photocurrent generated (A)
I _d	diode current (A)
Io	saturation diode current (A)
I _c	current at cell level (A)
Isc	short circuit current (A)
I_{mp}	maximum power point current (A)
$K_{ heta}\left(heta ight)$	incident angle modifier (–)
k _B	Boltzmann constant (J/K)
'n	22 ass flow rate (kg/s)
$m_e c_e$	solar collector heat capacity $(J/(m^2 K))$
п	ideality factor $(-)$

Lidiation delivered by a parabolic trough reflector. Concentrating PV/T collector using lenses are also characterized, i.e. the Fresnel concentring system presented by Chemisana et al. [5].

Jie et al. [6] developed a transient method for a hybrid PV/T collector wall (Building Integrated Photovoltaic–Thermal – BIPVT-type). The system was modeled and simulated with a special program, HYBRIDPV-1.0, written in FORTRAN.

A performance analysis of a PV/T collector by means of an explicit dynamic model was proposed by Chow [7]. The model is suitable for dynamic system simulation applications. The electrical characterization was conducted by using the electric power generated under similar terms as *the traditional linear expression* for the thermal dependence of the PV electrical efficiency [8]. Using the lineal expression for PV efficiency dependence, Solanki et al. [9] conducted an indoor characterization and modeling of a PV/T air collector concluding that the results obtained were similar with those obtained outdoors. In addition, the nature of the characteristic curve found out was similar to the one performed by Hottel–Whiller–Bliss for flat plate collector. A deep review about PV/T collectors was presented by Kumar and Rosen [10].

Zondag et al. [11] developed a simulation procedure for the thermal and electrical yield of a PV/T collector using a 3D dynamical model and three steady state models (3D, 2D and 1D). It was observed that the time-dependent model is required for an accurate prediction of the collector yield if the collector temperature at the end of a measurement differs from its starting temperature. The model proposed requires extensive knowledge of the *Runge-Kutta* numerical process. The *traditional linear expression* for PV electrical efficiency was included in the model as well.

Concerning the thermal performance, PV/T collectors have similar behavior with the conventional thermal modules. Since

N_p	number of lines of cells connected in parallel $(-)$
N _s	number of cells connected in series $(-)$
P_{th}	specific thermal power (W/m ²)
P_{el}	specific electrical power (W/m ²)
P_{mp}	maximum power (W)
r _s	series resistance (Ω)
r _{oc}	pen circuit resistance (Ω)
T _{in}	Lalet temperature (°C)
Tout	outlet temperature (°C)
T_m	mean fluid temperature (°C)
T_a	ambient temperature (°C)
T_c	cell temperature 🎇
U_L	overall convective eat loss coefficient of the collector
	$(W/(m^2 K))$
V	potential (V)
V_T	thermal potential (V)
V_c	potential at the cell level (V)
Voc	open circuit voltage (V)
Greek syi	mbols
η	cell/module electrical efficiency (%)
θ	angle of incidence (°)
τ_c	time constant (s)
τ_{lr}	linear model response time (s)
Δt_r	reading time interval (s)
Δt_s	sampling time interval (s)
Δt_f	filtering time interval (s)
$\varphi(T)$	thermal dependence (–)

not pecific standards exist, the official methodologies available for referring are the same than those for thermal collectors. In this way, discussion regarding steady-state, quasi-dynamic state and dynamic procedures is based on the one stated in a previous study of the authors [12]. This research analyses and proposes a simple dynamic linear model based on the piston flow concept. The main conclusions were that the time period for measurements should be at least one order of magnitude lower than the response time of the linear model. On the other hand, the outlet temperature and the corresponding power delivered by the collector can be calculated from a linear expression involving the average temporary integral of the weather variables. The averaging time must be equal to the linear model response time.

In terms of the electrical characterization, a derivation from the classical single diode model expression has been used instead of the basic and wide spread linear method to evaluate the maximum power output in operating conditions. The proposed method offers a much more detailed description of the module behavior under different working conditions, producing results for the characterization, independent of the maximum power point tracker used in the calibration. In the proposed methodology, the characteristic values of the different standards can be analytically obtained.

Considering the different aspects pointed out previously, the present research defines a PV/T characterization method based on the quasi-dynamic method [13] and the derived single diode model presented in [14–16]. In addition, the quasi-dynamic procedure is complemented taking into consideration the results obtained in Ref. [13]. The paper aims at obtaining a suitable PV/T model which better reflects the thermal and electrical behavior of the collector under stationary and transient conditions. The results are presented in two parts: (i) theoretical, (ii) experimental.

The theoretical part, Section 2, deals with the model definition and filtering method description. The experimental part includes Sections 3 and 4. The first one concerns the thermal and electrical results and the second one is the discussion of these results. Finally, the main conclusions are stated.

2. Theoretical background

2.1. The model

The thermal model used in this work is based on the simple dynamical equation presented in the EN 12975 norm (Eq. (H.2)). In order to adapt the model to hybrid PV/T collectors, the solar radiation term has been modified by the seful thermal radiation. The useful thermal radiation, G^* , is defined as the effective solar radiation which the collector converts into thermal power. Therefore, the radiation fraction transformed into electrical specific power (P_{el}) by the PV cells is subtracted to the total solar radiation: $G^* = G - P_{el.}$ Thus, the useful thermal specific power is expressed as:

$$P_{th} = F'K_{\theta}(\theta)(\tau\alpha)_{en}G^{*} - c_{1}(T_{m} - T_{a}) - c_{2}(T_{m} - T_{a})^{2} - c_{5}(dT_{m}/dt)$$
(1)

The useful thermal power, P_{th} , can also be approximated by the following equation:

$$P_{th} = \dot{m}c_p (T_{out} - T_{in})/A_{col}$$
⁽²⁾

where n^{15} the mass flow rate of fluid, c_p is the fluid heat capacity, $(T_{out} - T_{in})$ is temperature difference between fluid outlet and inlet, and A_{col} is the absorber plate area of the collector.

At this point it should be noted that according to other studies, for example Ref. [1], it is assumed that initially all the solar radiation is converted into thermal energy to obtain the electrical energy fraction. Thus, the useful thermal energy is derived. This treatment does not reflect the physical reality of the system and has a serious drawback: the photovoltaic cell temperature determined is much higher than the real one.

On the other hand, the electrical model employed regards the one stated by Chemisana and Rosell [16]. The useful specific electrical module power is described as:

$$P_{el} = \frac{N_p N_s}{A_{col}} V_c (I_{ph} + I_d + I_{sh})$$
(3)

The model consider 4 combination losses in the neutral and depletion zones together within the single diode model (I_d) an ideal current source (I_{ph}) to implement the photogenerated current and the current losses I_{sh} complete the model. The diode current equation is:

$$I_d = I_0 \left[\exp\left(\frac{V}{nV_T}\right) - 1 \right] \tag{4}$$

Che electrical parameterization introduces explicitly the irradiance and temperature dependence of some magnitudes: I_{ph} , I_0 and nV_T . The model assumes that the photocurrent generated I_{ph} is proportiona. The incident irradiance G and the reverse saturation diode current I_0 has a thermal dependence of $\varphi(T)$, where $\varphi(T) = T_c^3 \exp(-E_g/k_B T_c)$ is the factor depending on the band gap energy $E_{\rm g}$ of the cell constituent material, which is considered constant. T_c is the junction temperature (in K) and k_B is the Boltzman constant. 2 o temperature dependence of E_g , I_{ph} and I_{sh} is considered.

The detailed definition of each term used in the model is as follows:

$$I_{ph} = d_1 K_{\theta}(\theta)(\tau \alpha)_{en} G$$
⁽⁵⁾

$$I_d = d_2 \varphi(T) \left[\exp\left(\frac{(V_c + r_s I_c)}{n V_T}\right) - 1 \right]$$
(6)

$$I_{sh} = d_3(V_c + r_s I_c) \tag{7}$$

where V_c and I_c are the potential and the current at cell level calculated from the values at the module level in terms α_{11}^{11} is number of cells connected in series, N_s, and the number cell lines connected in parallel, N_n as:

$$V_c = \frac{V}{N_s} \tag{8}$$

$$I_c = \frac{I}{N_p} \tag{9}$$

The specific electrical power stated in Eq. (3), can be approached by:

$$P_{el} = \frac{VI}{A_{col}} \tag{10}$$

In the present research, it is assumed that the cell temperature is the same as the absorber plate temperature. Therefore, it is possible to obtain the cell temperature from the collector parameters as.

$$T_c = T_{in} + \frac{P_{th}}{F'U_L}(1 - F') \tag{11}$$

All these parameters are derived by applying the multiple linear regression process (MLR) on the measured data. This methodology is further described in the next sections.

2.2. Data acquisition and filtering methods

In the data acquisition and treatment process act three different time intervals:

- (1) The reading or capturing time interval Δt_r . It is the time when the system collects one data of each measured variable.
- (2) The sampling or measuring interval time Δt_s . It is the interval time when the system saves one averaged value from a certain measured data captured at Δt_r .
- (3) The filtering time interval Δt_{f} . It is the interval time used to calculate the mean in the moving average technique to reduce high frequency noise. The EU 12975 norm specifies that the reading time interval has to be from 1 to 10 s and the measuring time between 5 and 10 min. There are no other comments in the norm about the filtering interval time and h¹⁹ an be seen that the filtering time is equal to the measuring time.

Regarding the filtering, the Nyquist frequency f_N is half the sampling frequency (inverse of the sampling time interval) of a discrete signal processing system. The sampling theorem [17] shows that aliasing can be avoided if the Nyquist frequency is greater than the maximum component of the sampling frequency of the signal. In ³⁰ le output temperature of a solar collector the maximum frequency is related with the time constant of the collector $f_{\text{max}} = 1/\tau_c$, this means that the Nyquist frequency should be at least equal to this value. In consequence, the sampling frequency ($f_s = 1/\Delta t_s$) should be equal or higher than twice this value. The expression in the time domain is:

$$\Delta t_s \le \frac{\tau_c}{2} \tag{12}$$

And if it is assigned, taking a confidence range $\Delta t_s = \tau_c/10$ this condition fulfils Eq. (12).

In reference [12] was presented the piston flow model used to simulate the thermal behavior of a collector under dynamical conditions. The results of this research pointed out a linear algebraic expression to calculat²⁵he output temperature of the collector as a function of the average value of the radiation and the ambient temperature extended over the response time of the linear model.

The response time of the linear model was defined as:

$$\tau_{lr} = (1 - e^{-1})^{-1} \tau_c \tag{13}$$

where τ_c is the collector time constant which defined as the elapsed time between unshielding the collector to the sun and the point where the collector outlet temperature increases to $1 - e^{-1}$ (63.2%) from the initial value.

The mean radiation and ambient temperature terms could be understood as the backward moving average for the solar irradiance and the ambient temperature over the response time of the linear model.

In data processing, moving average is a technique used to reduce signal noise. It is used as a low pass filter to eliminate high frequency signal. The cut frequency $(f_c)^{26}$ defined as the inverse of the time length used in the average, according to Fourier Analysis. In this case, $f_c = 1/(\tau_{ir})$. Hence, signals with frequency higher than the cut frequency are eliminated.

This is the same effect that the collector makes over the meteorological data. If one variable has a change with a frequency higher than the related with the response time of the linear model, in the output the signal is filtered by the dissipative behavior of the thermal effect.

The result of the piston flow model points out that the filtering system based on moving average backwards, extended over a time period equal the response time of the linear model, is the most appropriated technique that can be used as a filtering process.

3. The experimental set-up

The tests were performed in a hybrid PV/T solar collecto³, the Applied Energy Research Centre (CREA) at the University of Lleida (Spain) which is located in Lleida at latitude 41.36°N and longitude 0.37°E. The data were collected during outdoor tests in November 2011.

The PV/T module consists o¹⁰ lines of 26 series connected crystalline silicon alls optimized for concentrating systems as can be seen in Fig. 1. The cells (NaPEC[®]) are attached to the heat sink, the adhesive used is a materia 7 ith high heat transfer conductivity that is also resistant to extreme temperatures and is an excellent electrical insulator (Chomerics Thermattach T404). The hybrid absorber is enclosed in an aluminium case with U cross-section; the encapsulation of the collector is finished fixing one top glass cover of 4 mm thickness. The module allows operating under vacuum conditions up to 10³ Pa to improve thermal efficiency. For the present study, the collector was investigated under normal radiation and atmospheric pressure. The main dimensions and characteristics of the collector are summarized in Table 1.

All the instruments and sensors were connected to datalogger CR23X³he inlet and outlet fluid and ambient temperatures, wind speed and solar radiation data were measured using type-K thermocouples, a Vector A-100R cup anemometer and a Kipp & Zonen CMP 11 pyranometer, respectively. The mass flow rate of the working fluid was accurately regulated by using a precision peristaltic pump (Percom N-M) which pumped at a constant known flow rate of 3.17 g/s or 0.022 kg/m² s. In order to achieve variations in inlet temperatures during the test, thermostatic immersion circulator (VWR 1122R) capable of water heating with an accuracy of ± 0.1 °C was used. Measurements were taken for several allet and outlet fluid temperatures, ambient temperatures, wind speed and solar radiation value. The different input variables were measured every 1 s and their mean values were recorded every 10 s. In paralleland thus, under the same weather conditions, the electrical dat.³ ere collected employing the IV tracer PVPM 2540C. The system was programmed to collect an IV characteristic curve every 10 min.

In order to have a good comparison with the steady-state results, the dynamic measurements were taken always with the sun perpendicular to the collector. This condition is possible to be reached due to the short measuring time period of time measurements needed as well as the use of a solar tracking system. The data were taken for a fixed low mass flow rate to increase that transfer process due to the small surface area of the PV/ T collector (Table 1). The same mass flow rate which was applied was kept constant during all the measurement tests.

The thermal experimental procedure conducted, was almost identical to that proposed in the EN. The only difference lies in



ig. 1. Picture of the PV/T collectors used in the present work, (b) the connection detail.

Table 1Data specification of the collectors.

Variable	Value
Mass flow rate (g/s) Module area (m ²) Cell area (cm ²) Mumber of cells connected in series (N_s) Number of cell lines connected in parallel (N_p) Volume (1) Residence time (s) Time constant (s)	3.17 0.146 16.8 52 1 0.071 22.0 87

the fact that the measurements were performed under forced transient conditions. A shading screen was used to fix a constant radiation value in a short time period, instead of measuring during the whole day. In Fig. 2 the collector behavior under these radiation constraints is shown. Two shading Lambertian diffusing screens were used: 1.5 m^2 each; transmittance coefficients: 54.0% and 72.5%, respectively.

3.1. Time constant measurement

The measuring of the time constant is needed in the characterization process. As described previously, the time constant determine time required for the outlet fluid temperature to attain ⁹3.2% of its steady-state value following a step change in the input (Fig. 3). The test method used is in agreement with the procedure described in the EN 12975 standard as presented in Ref. [12]. The final value of the time constant is the average of the values related with four cases with different measurement conditions. These measurements were initially conducted to determine the measured interval time. The measured interval time is defined as the bound period in which the calculations have to be performed. This period reflects the significant duration of the dynamic effect of the collector corresponding to the time derivative of the mean fluid temperature, which was correlated to $m_e c_e$ term in the quasi-dynamic model as given in Eq. (5).

⁸ ne effective thermal capacity m_ec_e and the time constant of the collector are essential parameters in the quasi-dynamic model which characterizes its transient performance. The effective thermal capacity m_ec_e can be considered as the sum, for each constituent element of the collector (glass, absorber, PV cells, liquid nationed, insulation), of the product of its mass, its specific heat and weighting factor. The m_ec_e can also be calculated from the energy balance of the collector as described in Ref. [13].

3.2. Steady-state and dynamical collector characterization

In terms of the dynamic thermal performance characterization, measurements were conducted for collecting data of inlet and out-



Fig. 2. (a) Representation of the experimental data set used to calibrate the model, (b) measured data set with applying moving average techniques. (Time in decaseconds, das)



Fig. 3. Time evolution of collector outlet fluid temperature T_{out} during an abrupt radiation change. The shape of the function permits to calculate collector time constant.

let fluid temperatures, ambient temperature, wind speed and solar radiation under forced transient conditions. As mentioned at the beginning of Section 3, in order to perform the measurement process, the collector was faced perpendicular to the sun at around solar noon. The inlet temperature of water was kept constant during five time intervals in the testing, whilst the other values were allowed to vary. Once both inlet and outlet fluid temperatures reached the steady-state conditions, the forced transient condition process was applied by suddenly covering the collector aperture area using the shading Lambertian screen. The collector was shielded for a period equal to four time constants, and then was exposed again to the sun until the next steady-state was achieved. Samples of the measured data are presented in Figs. 2 and 4.

According to the procedure described above, five test sequences were conducted to complete the rating of the collector. Each test corresponded to one over five different inlet fluid temperatures.

One test sequence of measurements has to be conducted ⁹ ith the inlet fluid temperature close to the ambient temperature in order to get the accurate value of zero loss efficiency parameter. The other four measurements have to be performed close to the middle and the highest of the temperature operation respectively. Meanwhile, each of the test sequences was divided into three steps: the pre-period to ensure that the steady-state was achieved, the forced transient period in which the forced conditions were applied through the shading screen and the last period to establish again the steady-state after removing the shading screen (Fig. 2). Each period took 3-4 times the time constant. Therefore, the total time needed for completing test sequence was around 9-12 time constants. Thus, the five tests for the five different inlet temperatures lasted for 27-36 the time constant plus the time required to increase the inlet temperature up to the next level. In this work the total time required was about 130 min.

3.3. Electrical measurements – reference characteristic IV curve

The electrical characteristic curve is very sensitive to both the temperature of the photovoltaic cell and the incident solar radiation. The temperature is related to the open-circuit voltage and the irradiance proportional to the short-circuit current (IEC 60904-5, IEC 60891) [18–19]. Accordingly, a reference characteristic IV curve is measured to calibrate the numerical model to refer the V_{oc} with the corresponding *T* and the I_{sc} with *G*.

In the case of the temperature, a problem arises in measuring the p-n junction temperature and to relate this with the correspondent open-circuit voltage. This is overcome thermalising the PV module at a stable temperature and protecting it from solar radiation. This process can be performed leaving the PV module in the morning to reach the ambient temperature at around the sunrise, when temperature has variations less than ± 1 °C/h. For the IV reference curve measurement, once the PV module is connected to the IV tracer, the procedure is as follows ¹¹he open-circuit voltage is measured whilst the module is covered, then the module is uncovered and an IV curve is collected, finally the module is covered again to take another V_{oc} value. These steps must be done as fast as possible to minimize the temperature effect in the measurement. The acquisition of the curve is considered correct if the open-circuit values taken at the beginning and at the end of the test achieve differences less than or equal to 5%.

The determination of the series resistance that appears in Eqs. (6) and (7) is performed according to the methodology presented by De Cueto [20]:

$$\left. \frac{dV}{dI} \right|_{OC} = \frac{n}{V_T} \frac{1}{I_{SC}} + r_s \tag{14}$$

The value of r_s is obtained by plotting the derivative of the characteristic curve at open circuit conditions in function of the inverse of the short circuit current. The independent term of the linear regression obtained is the r_s value, as it can be seen in Eq. (14). The rest of electrical parameters are obtained by MLR, as in the thermal characterization.

3.4. Parameters identification

In order to facilitate the parameter identification process, the data collection was simplified unplugging the photovoltaic module. Thus, the electrical production was negligible. There was only a small energy extraction when acquiring a characteristic curve about three seconds every 10 min. This simplification allows to firstly identify the thermal parameters. Then, the electrical parameters are determined.

During the experimental test, the data measurement was collected in order to estimate P_{th} values of the quasi-dynamic model. Multiple linear regression process and moving average (MA) method were applied to the data measurement in order to determine unknown parameter such as $F(\tau\alpha)_{en}$, c_1 , c_2 and c_5 as given in Eq. (1), where $F(\tau\alpha)_{en}$ is the series of ficiency, c_1 is the heat loss coefficient (FU_L), c_2 is the effective thermal capacity (m_ec_e) respectively.

The MLR method as a mathematical tool whicl.⁸ available in the most standard program with statistical function, was use.⁵ r the identification of the collector parameters. For an evaluation of the MLR process, all the *t*-ratio values of the parameters should be greater than the critical one. If this requirement is not fulfilled, the parameters which indicated with values of less than the critical value shoul.¹³ e set to zero and the parameter identification should be repeated with the adjusted collector model. The *t*-ratio is calculated a.⁵ he parameter value divided by the standard error of the same parameter.

As it can be seen in Figs. 2 and 4, the radiation fluctuations in the experimentally measured data obtained in this study were produced by the forced transient conditions. These variations appear jointly with the experimental system noise. The smoothing process is an important tool for reducing the effects of noise in time series. Therefore, the moving average method was applied to the measured data under forced transient conditions [21].

The length of the moving average filter is related with the linear model response time of the collector. As described in Section 2.2, in the present research the linear response time was fixed to be 1.58 collector time constants.



Fig. 4.² Time evolution of the measured and simulated output temperatures and the difference between them, (b) time evolution of the measured and simulated output temperatures with the use of moving average and the difference between them.

4. Results and discussion

4.1. Time constant and effective heat capacity

Using an initial Δt_s value of 1.0 s, a set of four determinations of the collector time constant corresponding to the four experimental measurements were performed. Fig. 3 shows one of these measurements. It can be observed the exponential behavior of the outlet temperature over time. It is important to notice that in these experiments the wate temperature was approximately equal to the ambient temperature while the collector was in thermal equilibrium with both the ambient and the inlet temperatures. The time constant was calculated using the methodology described in Section 3.1. The final value obtained was the mean of the four corresponding determinations, $\tau_c = 87 \pm 5s$. This result states that the use of a Δt_s of 10s is an adequate choice compatible with the specifications mentioned in paragraph 2.2. Consequently, $\Delta t_r = 1.0$ s was also used. This means that data collection was done capturing data every second and registering the average of the corresponding 10 lectures every 10 s in the dataloger memory.

The time constant also determines the linear model response time by means of Eq. (13) and in the present study takes the value of 140s. Therefore, this value has been used as the filtering time interval Δt_f in the backward moving average process.

In agreement with the procedure described in Section 3.1, four effective heat capacity values ($m_e c_e$) were calculated. The average value obtained was found to be 10.88 ± 0.19 kJ m⁻² K⁻¹.

4.2. Model characterization

The data presented in Fig. 3 were used for the collector thermal parameters characterization. These data were collected as described in the methodology discussed in Section 3.2 and by using the time intervals for the data collection obtained in the previous Section 4.1.

Table 2 sums up the steady-state results estimated from the linear regression applied to data obtained only for the intervals during which the inlet temperature remained constant (variation of inlet temperature less than ± 0.5 °C, as stated in the European Norm; in other words, for the 5 pleateau that can be identified in Fig. 2).

An essential validation for the experimental results was the compatibility among steady-state and dynamic. The values obtained by these two tests methods have to be compared within the same parameters. Therefore, the parameters of quasi-dynamic

 Table 2

 Results from the experimental tests for outdoor steady-state testing.

Parameters	Value ± SE (95% CI)	t-Ratio value	r ²
Steady-state $F'(\tau \alpha)_{en} (-)$ $F'U_L (W/(m^2K))$ $(m_e c_e) (kJ/(m^2 K))$	0.635 ± 0.001 15.2 ± 1.5 10.88 ± 0.19 ^a	635 9.99 55.77	0.9970

^a Value calculated from EN 12975 standard (Annex G).

method (Eq. (1)) are used and compared ⁵ ith the parameters of the steady-state model. The results obtained from the steady-state solution are closer to those of the dynamic-state one (Tables 2 and 3). It can also be seen that the coefficient of determination r^2 and the *t*-ratio values are quite reasonable for the two test methods. Based on these results, in can be noted that the forced transient conditions could give satisfactory results from three test sequences, which leads to a simple measuring procedure.

On other hand, regarding the forced transient conditions, MLR and MA methods applied in the current work, were acceptable for the dynamic model and for a short time period instead of taking measurements during the whole days. These measurements took much less total time than the measurements during the whole days as performed on the EN 12975 standard.

Tab	le	3
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Results from experimental tests for outdoor quasi-dynamic and dynamic testing.

Parameters	Value ± SE (95% CI)	t-ratio value	r ²
$\begin{array}{l} Quasi-dynamic \\ F' (\tau \alpha)_{en} (-) \\ F' U_L (W/(m^2 \mathrm{K})) \\ (m_e c_e) (\mathrm{kJ}/(m^2 \mathrm{K})) \end{array}$	0.610 ± 0.005 14.82 ± 0.21 6.8 ± 0.3	133.31 72.91 24.07	0.9735
$\begin{array}{l} Dynamic-state \ with \ MA\\ F'\ (\tau\alpha)_{en}\ (-)\\ F'U_L\ (W/(m^2\ K))\\ (m_ec_e)\ (kJ/(m^2\ K)) \end{array}$	$\begin{array}{c} 0.641 \pm 0.002 \\ 15.38 \pm 0.08 \\ 10.62 \pm 0.16 \end{array}$	337.27 190.12 67.69	0.9959

The coefficients for the quasi-dynamic characterization are the same with those for the dynamic one (Table 3), however the dynamic-state coefficients are calculated by using the moving average technique described in Section 2.2 (Eqs. (13) and (14)). The temporal filtering (MA) has been applied not only to the variables T_a and G, but it has been extended to the variables T_{in} and T_{out} (Fig. 4b). The time interval used in this filter was 140s, which corresponds to $\tau = (1-e^{-1})^{-1} \tau_c$. Another important modification with respect to the EN quasi-dynamic method is the following: for obtaining the above mentioned parameters, all the data recorded including the points with variable T_{in} were considered. The inclu-



Fig. 5. Measured outlet temperatures T_{out} vs. the modeled ones: (a) without applying the MA technique, (b) with applying the MA technique.



Fig. 6. Cell electrical resistance $R_{oc} (dV/dI)_{oc}$ vs. the inverse of the short circuit cell current $(1/I_{sc})$.

sion of these new items improved in all cases the statistical descriptors used for the evaluation of the methodology.

The most significant difference introduced by using the moving average technique is the stabilization of the obtained effective heat capacity value. As remarked in the introduction an ²⁹ ported in the literature, one of the major drawbacks of using the quasi-dynamic standard method is the variability of the values obtained by the linear regression. From the results presented in Tables 2 and 3, it can be noted that the numbers obtained through the proposed procedure are the closest determinations achieved in reference to the measurements collected. The relative discrepancy of the effective heat capacity with respect to the experimental data was 62.9% for the standard method and 1.5% for the proposed method.

After adjusting the parameter in the identification process, all the *t*-ratio have an absolute value greater than the critical one, therefore this means that all the parameters in the regression equation are useful in predicting the assessed value of the quasidynamic model.

The results calculated for the outlet temperature are plotted in Fig. 4a for the standard method and in Fig. 4b for the proposed

method using the moving average technique. In both cases the simulated temperature was calculated using an iterative process (Eq. (1)) and setting $P_{el} = 0$. In Fig. 5 the comparison between the outlet experimental temperatures and modeled ones obtained from both methods, is depicted. The moving average provides a better estimate than the original values because noise is reduced. The measured data fluctuations were smoothed to expose its features and provide a reasonable starting approach for parametric fitting.

Fig. 7 plots a set of the most representative experimental I-V curves corresponding to different irradiances and temperatures (circles). Each curve is represented by 100 measured points.

From the experimental *I*–*V* values and following the methodology described in Section 3.4, the resistance under open circuit conditions r_{oc} was calculated. Fig. 6 shows the correlation between the resistances r_{oc} and the inverse short circuit currents obtained under the same weather conditions. In Fig. 6, the linear equation which best correlates both variables, is also presented. The y-interception of the equation found, according to Eq. (14), represents the value of the mean series resistance r_s of an average cell of the module. In this case $r_s = 0.7 \text{ m}\Omega$.

The constants d_1 , d_2 and d_3 were determined by fitting Eq. (2) to the experimental data. The results of the three constants, a_{28}^{28} are series resistance and the ideality factor are given in Table 4. The goodness of the fit can be appreciated graphically in Fig. 7, where a set of measured curves is plotted 6 gainst the simulated values obtained by substituting the parameters from Table 4 in Eq. (2). Using the same temperature and irradiance values, the slope obtained was almost 1 while the coefficient of determination was 0.999 (Fig. 8). From the *t*-Student parameter analysis, all the parameters achieve a *t*-value above the critical so they have significance in defining the final value.

Fig. 9 shows the comparison between the average cell temperature determined experimentally and the one calculated by Eq. (10). In Fig. 9, the results from applying the same equation to particular input values, such as the mean fluid, the outlet and the inlet temperatures. This figure demonstrates that the inlet temperature is the one which best adjusts the cell temperature.



Fig. 7. Set of experimental and simulated representative electrical IV characteristic curves rolar radiation G ranging from 500 to 1050 W m⁻² and cell temperature T_c ranging from 17 to 45 °C. This set of data was used to derive the electrical parameters from the model.

Table 4Cell electrical coefficients.

	$d_1 ({\rm Am}^2{\rm W}^{-1})$	$d_2 (AK^{-3})$	$d_3\left(\Omega^{-1} ight)$	n (-)	$r_{s}\left(\Omega ight)$
Value Standard error	$\begin{array}{c} 5.14\times 10^{-4} \\ 1.41\times 10^{-6} \end{array}$	-731.02 2.41	$\begin{array}{c} -9.75\times 10^{-3} \\ 3.24\times 10^{-3} \end{array}$	1.19	7.45×10^{-4}



Fig. 8. Correlation between the simulated and experimental currents. The simulated values were obtained from Eqs. (2)–(8) and by using the coefficients given in Table 4.



Fig. 9. Results for the cell temperature obtained by Eq. (1²)ing different values of input temperature compared with the experimental values.

Table 5 Electrical characteristics at standard conditions (25 $^\circ C$ and 1.0 $kWm^{-2}).$

	Average cell	Module
V_{oc} (V)	0.586	30.51
$I_{sc}(A)$	0.510	0.510
$P_{mp}(W)$	0.241	12.53
V_{mp} (V)	0.500	26.00
I_{mp} (A)	0.483	0.483
η (%)	14.3	8.55
FF (-)	0.802	0.802

It is noteworthy that the electrical characterization performed provides a good description of the collector behavior as photovoltaic module for any working condition. This amplifies significantly the photovoltaic characterization methodology defined under the standard IEC 60904 [18]. As described in the standard, the characterization is carried out only under determined environmental conditions (standard conditions STC). These conditions are extended approximately to other conditions by using the T_{NOC} equation and by introducing other thermal parameters such as alpha and beta [22], sometimes not measured directly on the module,

depending on its manufacturing technology. Finally, the main electrical characteristics for the cell and for the module are given in Table 5. In the efficiency calculation, the PV cell coverage ratio obtained by dividing the cell area by the module area (0.598), is used to relate both the cell and the module values [23].

5. Conclusions

In this paper, a modified test to characterize the behavior of a PVT solar collector was developed and carried out. The model presente an extension and coupling of the quasi-dynamic thermal test with the single-diode photovoltaic model. The joining term is the effective solar radiation delivered on the thermal absorber.

The forced transient conditions were applied to the incoming solar radiation in order to increase the thermal capacitance effect and the thermal variability. This led to a simpler process and a shorter test period for quasi-dynamic performance tests. The forced conditions were achieved through the use of a Lambertian screen.

An adequate sampling interval time and the backward moving average technique used allow reduction of the heat capacity fluctuations, and thus the linear regression results are improved considerably. Besides, this permits to model the system with no constant inlet temperature requirements. The sampling interval time was set to be approximately 1/10 the time constant. The filter length used in the moving average was fixed to 1.58 times the collector time constant.

²he electrical parameterization introduced explicitly the irradiance dependence of the photogenerated current and the thermal dependence of the reverse saturation diode current. Contrarily, no temperature dependence in the gap energy and photogenerated current was considered.

⁶he electrical performance had been analyzed through the *I–V* curves measured at different temperature and illumination outdoor conditions. By modeling these data, the electrical parameters were used to indicate the solar radiation and the cell temperature. ²he temperature of the cell had been related with the temperature of the absorber using a theoretical expression. The cell temperature can be determined by the inlet temperature and the effective thermal specific power of the collector. The model results fit well with the experimental values, achieving a correlation coefficient of 0.999.

The series resistance was estimated correlating the open circuit resistance against the inverse short circuit current. The average cell series resistance corresponds to the y-interception of the correlation equation.

The model was applied to characterize a PV/T collector. The zero efficiency and heat loss values obtained after applying this process showed a good degree of agreement with the results obtained from the steady-state thermal performance test.

The methodology proposed for the electrical characterization presents more flexibility to obtain the output power for wide range of weather conditions than the model stated in the standard IEC 61215 used for the characterization and certification of photovoltaic modules for terrestrial applications. The proposed model reproduces the photovoltaic module behavior under much more variable working conditions than the international standard and extends the final result application field.

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